



Description

The AL8861 is a hysteresis mode DC-DC step-down converter, designed for driving single or multiple series connected LEDs efficiently from a voltage source higher than the LED voltage. The device can operate from an input supply between 4.5V and 40V and provide an externally adjustable output current up to 1A for TSOT25 package and 1.5A for SOT89-5 and MSOP-8EP packages. Depending upon supply voltage and external components, this converter can provide up to 40W of output power.

The AL8861 integrates the power switch and a high-side output current sensing circuit, which uses an external resistor to set the nominal average output current.

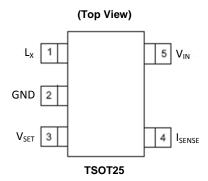
Dimming can be realized by applying an external control signal to the V_{SET} Pin. The V_{SET} Pin will accept either a DC voltage signal or a PWM signal.

The soft-start time can be adjusted by an external capacitor from the V_{SET} Pin to Ground. Applying a voltage of 0.2V or lower to the V_{SET} Pin can turn off the output and make the device enter into standby state with low current consumption.

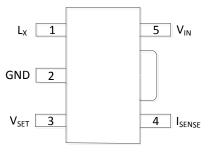
Features

- Low BOM Counts
- Wide Input Voltage Range: 4.5V to 40V
- Output Current up to 1.5A
- Internal 40V NDMOS Switch
- Typical 5% Output Current Accuracy
- Single Pin for On/Off and Brightness Control by DC Voltage or PWM Signal
- Recommended Analog Dimming Range: 5% to 100%
- Soft-Start
- High Efficiency (Up to 97%)
- LED Short Protection
- Inherent Open-Circuit LED Protection
- R_S Short Protection
- Over Temperature Protection (OTP)
- Up to 1MHz Switching Frequency
- Pb-free TSOT25, SOT89-5 and MSOP-8EP Packages
- Totally Lead-Free & Fully RoHS Compliant (Notes 1 & 2)
- Halogen and Antimony Free. "Green" Device (Note 3)

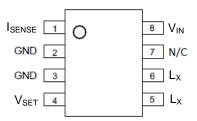
Pin Assignments



(Top View)



SOT89-5



MSOP-8EP

Applications

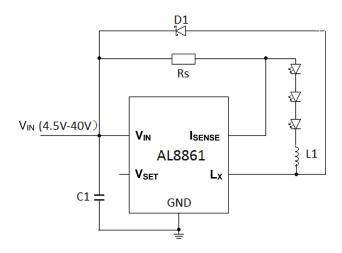
- LED Retrofit for Low Voltage Halogen
- Low Voltage Industrial Lighting
- LED Backlighting
- Illuminated Signs
- External Driver with Multiple Channels

Notes:

- 1. No purposely added lead. Fully EU Directive 2002/95/EC (RoHS) & 2011/65/EU (RoHS 2) compliant.
- 2. See http://www.diodes.com/quality/lead_free.html for more information about Diodes Incorporated's definitions of Halogen- and Antimony-free, "Green" and Lead-free.
- 3. Halogen- and Antimony-free "Green" products are defined as those which contain <900ppm bromine, <900ppm chlorine (<1500ppm total Br + Cl) and <1000ppm antimony compounds.



Typical Applications Circuit

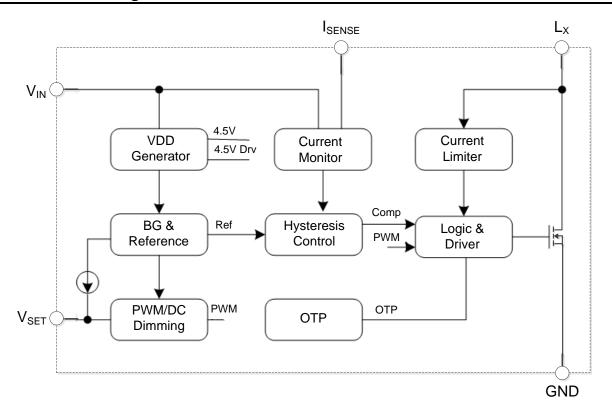


Pin Descriptions

Pin N	Pin Number			
TSOT25/ SOT89-5	MSOP-8EP	Pin Name	Function	
1	5, 6	L _X	Drain of NDMOS switch.	
2	2, 3	GND	Ground (0V)	
3	4	V _{SET}	 Multi-Function On/Off and Brightness Control Pin: Leave floating for normal operation Drive to voltage below 0.2V to turn off output current Drive with DC voltage (0.3V < V_{SET} < 2.5V) to adjust output current from 0 to 100% of I_{OUTNOM}. Linear adjustment range from 5% to 100% of I_{OUTNOM} Drive with PWM signal from open-collector or open-drain transistor, to adjust output current. Linear adjustment range from 1% to 100% of I_{OUTNOM} for f < 500Hz Connect a capacitor from this pin to Ground to increase soft-start time. (Default soft-start time = 0.1ms. Additional soft-start time is approximately 1.5ms/1nF)	
4	1	I _{SENSE}	Connect resistor R_S from this pin to V_{IN} to define nominal average output current. $I_{OUTNOM} = 0.1/R_S$	
5	8	VIN	Input voltage (4.5V to 40V). Decouple to Ground with $10\mu F$ or higher X7R ceramic capacitose to device.	
_	7	N/C	No connection	



Functional Block Diagram



Absolute Maximum Ratings (Note 4)

Symbol	Parameter Rating		9	Unit
V _{IN}	Input Voltage	-0.3 to +	-0.3 to +42	
V _{LX} ,V _{ISENSE}	L _X , I _{SENSE} Pin Voltage	-0.3 to +	-0.3 to +42	
V _{VSET}	V _{SET} Pin Voltage	-0.3 to -	+6	V
TJ	Operating Junction Temperature	+150		°C
T _{STG}	T _{STG} Storage Temperature Range		-65 to +150	
T _{LEAD}	Lead Temperature (Soldering, 10sec)	+260		°C
		TSOT25 (Note 5)	147	
θ_{JA}	Thermal Resistance (Junction to Ambient)	SOT89-5 (Note 6)	116	°C/W
		MSOP-8EP (Note 7)	56	
		TSOT25 (Note 5)	27	
θ_{JC}	Thermal Resistance (Junction to Case)	SOT89-5 (Note 6)	24	°C/W
		MSOP-8EP (Note 7)	15	

Note:

- 4. Stresses greater than those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "Recommended Operating Conditions" is not implied. Exposure to "Absolute Maximum Ratings" for extended periods may affect device reliability.
- $5.\ Device\ mounted\ on\ 1"x1"\ FR-4\ MRP\ substrate\ PC\ board,\ 2oz\ cooper,\ with\ minimum\ recommended\ pad\ layout.$
- 6. Device mounted on 1"*1" FR-4 substrate PC board, 2oz copper, with minimum recommended pad layout.
- 7. Device mounted on 2"*2" FR-4 substrate PC board, 2oz copper, with minimum recommended pad layout.



Recommended Operating Conditions

Symbol	Parameter	Parameter		Max	Unit
V _{IN}	Input Voltage		4.5	40	٧
f _{SW}	Switching Frequency		_	1	MHz
	Continuous Output Current	TSOT25	_	1	А
Гоит		SOT89-5	_	1.5	
		MSOP-8EP	_	1.5	
T _A	Operating Ambient Temperature		-40	+85	°C
TJ	Operating Junction Temperature		-40	+125	°C
_	Recommended Analog Dimm	ing Range	5	100	%

Electrical Characteristics (@ $V_{IN} = 16V$, $\underline{T}_A = +25$ °C, unless otherwise specified.)

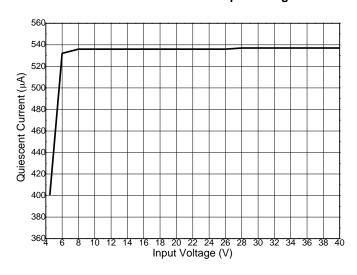
Symbol	Parameter	Conditions	Min	Тур	Max	Unit
SUPPLY VOLT	SUPPLY VOLTAGE					
V _{IN}	Input Voltage	_	4.5	_	40	V
IQ	Quiescent Current	V _{SET} Pin Floating, V _{IN} = 16V		0.55	_	mA
I _{SHDN}	Shutdown Supply Current	V _{SET} Pin Grounded		55	100	μΑ
V _{SENSE}	Mean Current Sense Threshold Voltage	Measured on I _{SENSE} Pin with Respect to V _{IN}	96	100	104	mV
V _{SENSE_HYS}	Sense Threshold Hysteresis	_		±13	_	%
I _{SENSE}	I _{SENSE} Pin Input Current	V _{SENSE} = V _{IN} -0.1V		8	_	μΑ
V _{EN}	V _{SET} Range on V _{SET} Pin	For Analog Dimming	0.3	_	2.5	V
V _{EN(ON)}	DC Voltage on V _{SET} Pin to Enable	VEN Rising	_	0.25	_	V
V _{EN(OFF)}	DC Voltage on V _{SET} Pin to Disable	Ven Falling	_	0.2	_	V
R _{LX}	L _X Switch On-Resistance	@I _{LX} = 100mA	_	0.2	_	Ω
I _{LX(LEAK)}	L _X Switch Leakage Current	_	_	_	5	μA
t _{SS}	Soft-Start Time	V _{IN} = 16V, C _{EN} = 1nF	_	1.5	_	ms
f _{LX}	Operating Frequency	$V_I = 16V, V_O = 9.6V (3 LEDs)$ L = 47µF, $\Delta I = 0.25A (I_{LED} = 1A)$	_	250	_	kHz
t _{ON_REC}	Recommended Minimum Switch ON Time	For 4% Accuracy	_	500	_	ns
f _{LX(MAX)}	Recommended Maximum Switch Frequency	_		_	1.0	MHz
D _{LX(MAX)}	Max Duty Circle	_	_	98	_	%
D _{LX}	Recommended Duty Cycle Range	_	25	_	75	%
t _{PD}	Internal Comparator Propagation Delay (Note 8)	_		45	_	ns
T _{OTP}	Overtemperature Protection	_	_	+150	_	°C
T _{OTP_HYS}	Temp Protection Hysteresis	_	_	+30	_	°C
I _{XL(MAX)}	Current Limit	Peak Inductor Current	2	_	_	Α

Note: 8. Guaranteed by design.

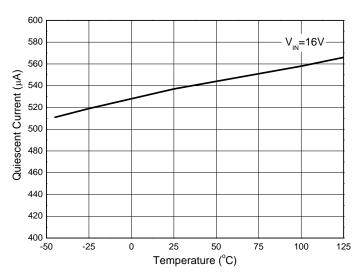


Typical Performance Characteristics ($@T_A = +25^{\circ}C$, $V_{IN} = 16V$, unless otherwise specified.)

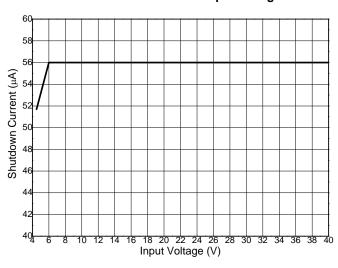
Quiescent Current vs. Input Voltage



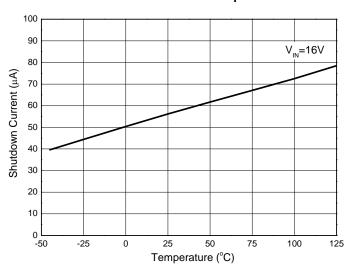
Quiescent Current vs. Temperature



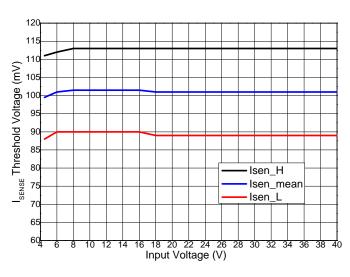
Shutdown Current vs. Input Voltage



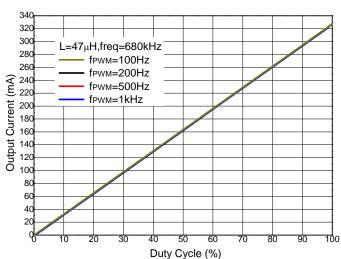
Shutdown Current vs. Temperature



I_{SENSE} Threshold Voltage vs. Input Voltage

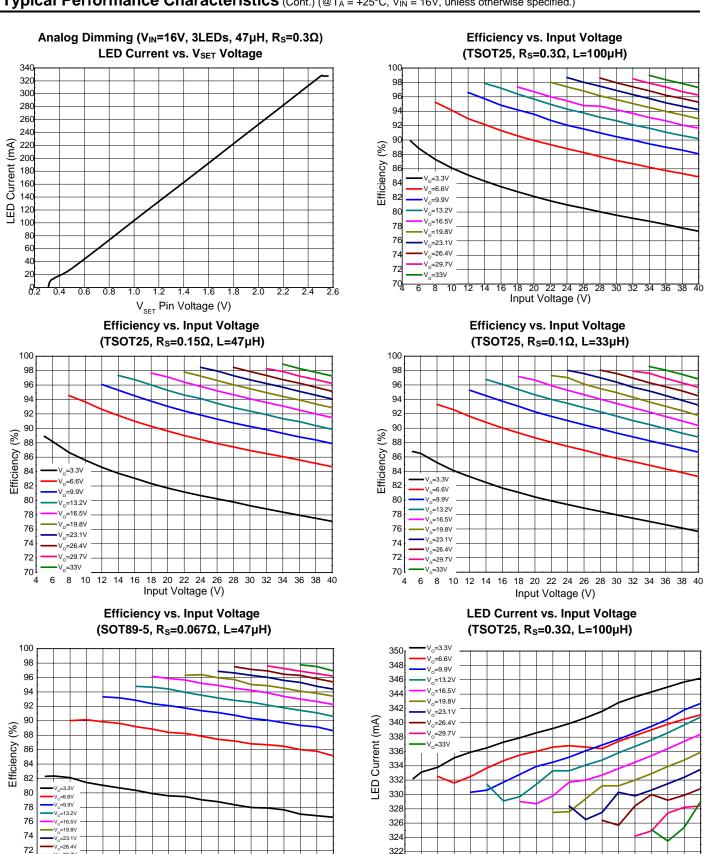


PWM Dimming (V_{IN} =16V, 3 LEDs, 47 μ H, R_S=0.3 Ω) LED Current vs. Duty Cycle





Typical Performance Characteristics (Cont.) (@T_A = +25°C, V_{IN} = 16V, unless otherwise specified.)



8

10 12 14 16 18 20 22 24 26 28 30 32 34 36 38 40

Input Voltage (V)

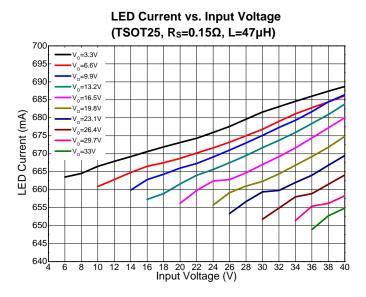
70

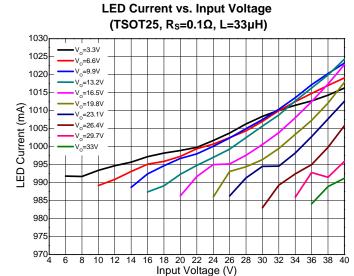
8 10 12 14 16 18 20 22 24 26 28 30 32 34 36 38 40

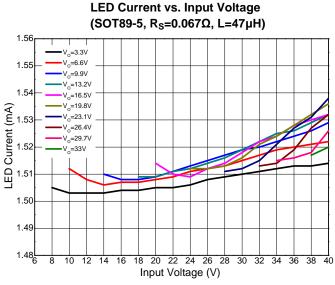
Input Voltage (V)

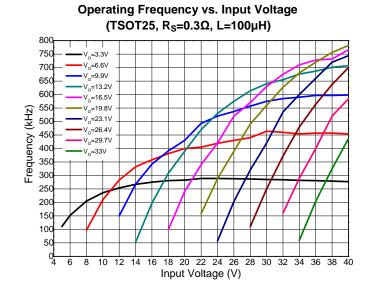


Typical Performance Characteristics (Cont.) (@T_A = +25°C, V_{IN} = 16V, unless otherwise specified.)

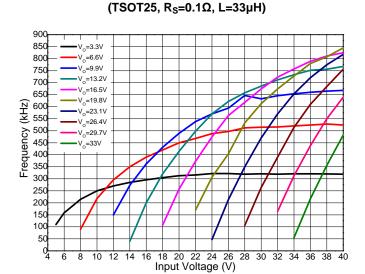




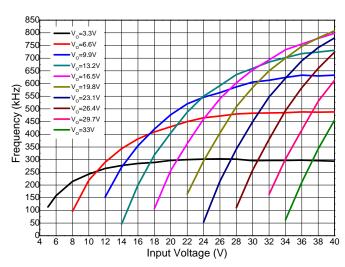




Operating Frequency vs. Input Voltage (TSOT25, R_S=0.15Ω, L=47μH)

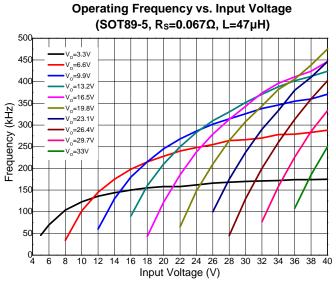


Operating Frequency vs. Input Voltage

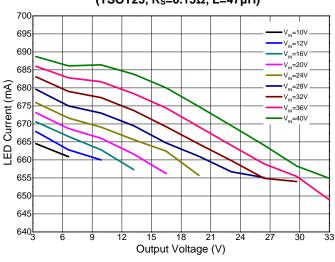




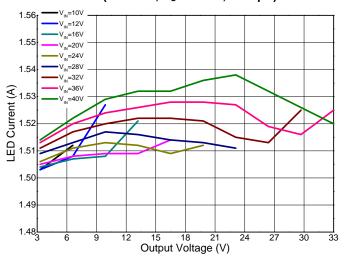
Typical Performance Characteristics (Cont.) (@T_A = +25°C, V_{IN} = 16V, unless otherwise specified.)



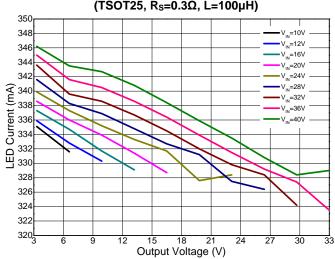
LED Current vs. Output Voltage (TSOT25, $R_S=0.15\Omega$, L=47 μ H)



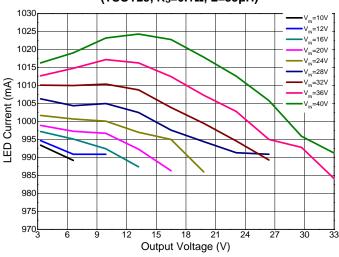
LED Current vs. Output Voltage (SOT89-5, $R_s=0.067\Omega$, L=47 μ H)



LED Current vs. Output Voltage (TSOT25, R_S=0.3Ω, L=100μH)



LED Current vs. Output Voltage (TSOT25, R_S=0.1Ω, L=33μH)





I_L 100mA/div

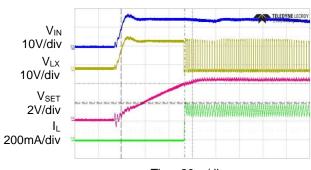
 $\begin{array}{c} V_{LX} \\ 5V/div \end{array}$

Performance Characteristics (@V_{IN} = 16V, 3 LEDs, R_S = 0.3Ω , L = 47μ H, T_A = $+25^{\circ}$ C, unless otherwise specified.)

Steady State TELEDYNE LECROY Everywhereyoulcok

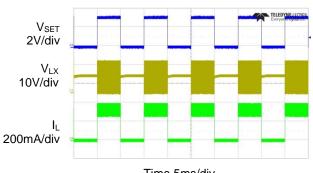
Time 1µs/div

Start Up



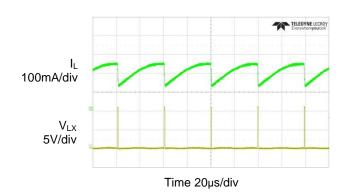
Time 20µs/div

PWM Dimming (100Hz, Duty=50%)

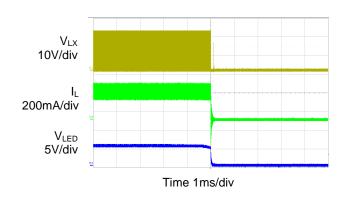


Time 5ms/div

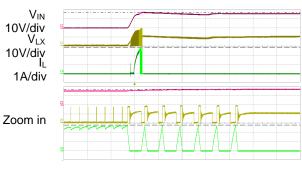
Pulse Skip Mode



LED Open Protection



R_S Short Protection



Time 5ms/div



Application Information

The AL8861 is a hysteretic mode LED driver with integrated power switch. It is available in two packages that provide a trade-off between PCB area and power dissipation capability. It is recommended that at higher LED currents/smaller PCBs that the SOT89-5 version should be used to maximize the allowable LED current over a wider ambient temperature range.

AL8861 Operation

In normal operation, when normal input voltage is applied at $+V_{IN}$, the AL8861 internal switch will turn on. Current starts to flow through sense resistor R_S , inductor L1, and the LEDs. The current ramps up linearly, and the ramp rate is determined by the input voltage V_{IN} and the inductor L1. This rising current produces a voltage ramp across R_S . The internal circuit of the AL8861 senses the voltage across R_S and applies a proportional voltage to the input of the internal comparator. When this voltage reaches an internally set upper threshold, the internal switch is turned off. The inductor current continues to flow through R_S , L1, LEDs and diode D1, and back to the supply rail, but it decays, with the rate determined by the forward voltage drop of LEDs and the diode D1.

This decaying current produces a falling voltage on R_S, which is sensed by the AL8861. A voltage proportional to the sense voltage across R_S will be applied at the input of internal comparator. When this voltage falls to the internally set lower threshold, the internal switch is turned on again.

This switch-on-and-off cycle continues to provide the average LED current set by the sense resistor Rs.

LED Current Configuration

The nominal average output current in the LED(s) is determined by the value of the external current sense resistor (R_S) connected between V_{IN} and I_{SENSE} and is given by:

$$I_{OUT(NOM)} = \frac{0.1}{Rs}$$

The table below gives values of nominal average output current for several preferred values of current setting resistor (R_S) in the Typical Application Circuit shown on Page 2.

R _S (Ω)	Nominal Average Output Current (mA)		
0.066	1,500		
0.1	1,000		
0.13	760		
0.15	667		
0.3	333		

The above values assume that the V_{SET} Pin is floating and at a nominal reference voltage for internal comparator. It is possible to use different values of R_S if the V_{SET} Pin is driven by an external dimming signal.

Analog Dimming

Applying a DC voltage from 0.3V to 2.5V on the V_{SET} Pin can adjust output current from 0 to 100% of I_{OUTNOM} , as shown in Figure 1. Recommended dimming range is from 5% to 100%. If the V_{SET} Pin is brought higher than 2.5V, the LED current will be clamped to 100% of I_{OUTNOM} while if the V_{SET} voltage falls below the threshold of 0.3V, the output switch will turn off.

PWM Dimming

LED current can be adjusted digitally, by applying a low frequency pulse-width-modulated (PWM) logic signal to the V_{SET} Pin to turn the device on and off. This will produce an average output current proportional to the duty cycle of the control signal. To achieve a high resolution, the PWM frequency is recommended to be lower than 500Hz, however higher dimming frequencies can be used at the expense of dimming dynamic range and accuracy. Typically, for a PWM frequency of 500Hz the accuracy is better than 1% for PWM ranging from 1% to 100%.

The accuracy of the low duty cycle dimming is affected by both the PWM frequency and the switching frequency of the AL8861. For best accuracy/resolution, the switching frequency should be increased while the PWM frequency should be reduced.

The V_{SET} Pin is designed to be driven by both 3.3V and 5V logic levels directly from a logic output with either an open drain output or push pull output stage.



Application Information (Cont.)

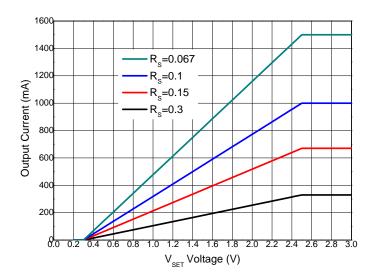


Figure 1. Analog Dimming Curve

Soft-Start

The default soft-start time for AL8861 is only 0.1ms - this provides very fast turn-on of the output, improving PWM dimming accuracy.

Nevertheless, adding an external capacitor from the V_{SET} Pin to Ground will provide a longer soft-start delay. This is achieved by increasing the time for the V_{SET} voltage rising to the turn-on threshold, and by slowing down the rising rate of the control voltage at the input of hysteresis comparator. The additional soft-start time is related to the capacitance between V_{SET} and GND, the typical value will be 1.5ms/nF.

Capacitor Selection

A low ESR capacitor should be used for input decoupling, as the ESR of this capacitor appears in series with the supply source impedance and will lower overall efficiency. This capacitor has to supply the relatively high peak current to the coil and smooth the ripple on the input current.

The minimum capacitance needed is determined by input power, cable's length and peak current. $4.7\mu\text{F}$ to $10\mu\text{F}$ is a commonly used value for most cases. A higher value will improve performance at lower input voltages, especially when the source impedance is high. The input capacitor should be placed as close as possible to the IC.

For maximum stability of overtemperature and voltage, capacitors with X7R, X5R or better dielectric are recommended. Capacitors with Y5V dielectric are not suitable for decoupling in this application and should NOT be used.

Diode Selection

For maximum efficiency and performance, the freewheeling diode (D1) should be a fast low capacitance Schottky diode with low reverse leakage current. It also provides better efficiency than silicon diodes, due to lower forward voltage and reduced recovery time.

It is important to select parts with a peak current rating above the peak coil current, and a continuous current rating higher than the maximum output load current. It is very important to consider the reverse leakage current of the diode when operating above +85°C. Excess leakage current will increase power dissipation.

The higher forward voltage and overshoot due to reverse recovery time in silicon diodes will increase the peak voltage on the L_X output. If a silicon diode is used, more care should be taken to ensure that the total voltage appearing on the L_X Pin including supply ripple, won't exceed the specified maximum value.



Application Information (Cont.)

Inductor Selection

Recommended inductor values for the AL8861 are in the range 33μ H to 100μ H. Higher inductance are recommended at higher supply voltages in order to minimize output current tolerance due to switching delays, which will result in increased ripple and lower efficiency. Higher inductance also results in a better line regulation. The inductor should be mounted as close to the device as possible with low resistance connections to L_X pins.

The chosen coil should have saturation current higher than the peak output current and a continuous current rating above the required mean output current.

The inductor value should be chosen to maintain operating duty cycle and switch 'on'/'off' times within the specified limits over the supply voltage and load current range. The following equations can be used as a guide.

$$t_{ON} = \frac{L\Delta I}{V_{IN} - V_{LED} - I_{LED}(R_S + R_L + R_{LX})}$$

$$t_{OFF} = rac{L_{ ext{X}} \, ext{Switch 'Off' Time}}{V_{LED} + V_D + I_{LED} (R_S + R_L)}$$

Where: L is the coil inductance; R_L is the coil resistance; R_S is the current sense resistance; I_{LED} is the required LED current; ΔI is the coil peak-peak ripple current (internally set to $0.26 \times I_{LED}$); V_{IN} is the supply voltage; V_{LED} is the total LED forward voltage; R_{LX} is the switch resistance (0.2Ω nominal); V_D is the diode forward voltage at the required load current.

Thermal Protection

The AL8861 includes Over-Temperature Protection (OTP) circuitry that will turn off the device if its junction temperature gets too high. This is to protect the device from excessive heat damage. The OTP circuitry includes thermal hysteresis that will cause the device to restart normal operation once its junction temperature has cooled down by approximately +30°C.

Thermal Fold-back

The AL8861 has thermal fold-back function: The IC has analog dimming function with a DC voltage from 0.3V to 2.5V on V_{SET} pin to adjust output current from 0 to 100% of I_{OUT_NOM} linearly. The V_{SET} pin allows the direct connection of an NTC to ground to sense ambient temperature. The IC has internal constant current source of 1.5µA to charge V_{SET} pin. When the ambient temperature increases, the V_{SET} pin voltage decreases which equals to 1.5µA x R_{NTC} . When the V_{SET} pin voltage drops below 2.5V, the output current is decreased proportionally to the V_{SET} pin voltage.

Open Circuit LEDs

The AL8861 has by default open LED protection. If the LEDs should become open circuit the AL8861 will stop oscillating; the I_{SENSE} pin will rise to V_{IN} and the I_{LX} pin will then fall to GND. No excessive voltages will be seen by the AL8861.

LED Chain Shorted Together

If the LED chain should become shorted together (the anode of the top LED becomes shorted to the cathode of the bottom LED) the AL8861 will continue to switch and the current through the AL8861's internal switch will still be at the expected current - so no excessive heat will be generated within the AL8861. However, the duty cycle at which it operates will change dramatically and the switching frequency will most likely decrease. See Figure 2 for an example of this behavior at 24V input voltage driving 3 LEDs.

The on-time of the internal power MOSFET switch is significantly reduced because almost all of the input voltage is now developed across the inductor. The off-time is significantly increased because the reverse voltage across the inductor is now just the Schottky diode voltage (See Figure 2) causing a much slower decay in inductor current.



Application Information (Cont.)

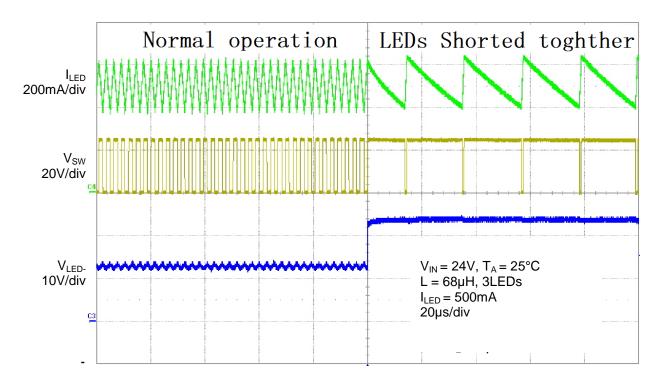


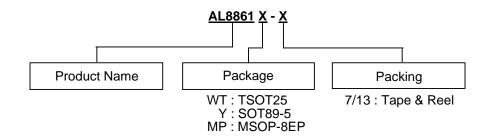
Figure 2. Switching Characteristics (Normal Operation to LED Chain Shorted Out)

R_S Short Protection

The AL8861 has a current limit at about 2.8A. If R_S is shorted, current limit is triggered for accumulated 7 times and the switch will shut down and latch up.



Ordering Information (Note 9)

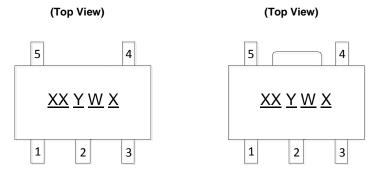


David November	Deales va Oada	Dealess	Tape and Reel		
Part Number Package Code		Package	Quantity	Part Number Suffix	
AL8861WT-7	WT	TSOT25	3,000/Tape & Reel	-7	
AL8861Y-13	Υ	SOT89-5	2,500/Tape & Reel	-13	
AL8861MP-13	MP	MSOP-8EP	2500/Tape & Reel	-13	

Note: 9. For packaging details, go to our website at http://www.diodes.com/products/packages.html.

Marking Information

(1) TSOT25, SOT89-5



TSOT25 SOT89-5

XX: Identification Code

Y: Year 0 to 9

<u>W</u>: Week: A to Z: 1 to 26 week;

a to z : 27 to 52 week; z represents

52 and 53 week

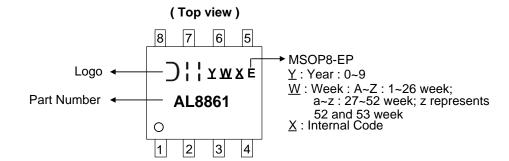
X: Internal Code

Part Number	Package	Identification Code
AL8861WT-7	TSOT25	A4
AL8861Y-13	SOT89-5	A4



Marking Information (Cont.)

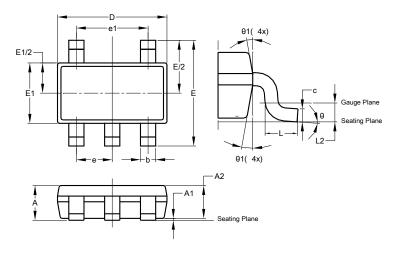
(2) MSOP-8EP



Package Outline Dimensions

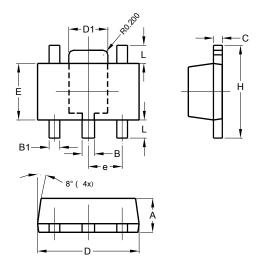
Please see http://www.diodes.com/package-outlines.html for the latest version.

(1) Package Type: TSOT25



TSOT25				
Dim	Min	Max	Тур	
Α	-	1.00	-	
A1	0.01	0.10	-	
A2	0.84	0.90	-	
b	0.30	0.45	-	
С	0.12	0.20	-	
D	-	-	2.90	
Е	-	-	2.80	
E1	-	-	1.60	
е	(0.95 BS	С	
e1		1.90 BS	С	
L	0.30	0.50		
L2	0.25 BSC			
θ	0°	8°	4°	
θ1	4°	12°	-	
All [All Dimensions in mm			

(2) SOT89-5



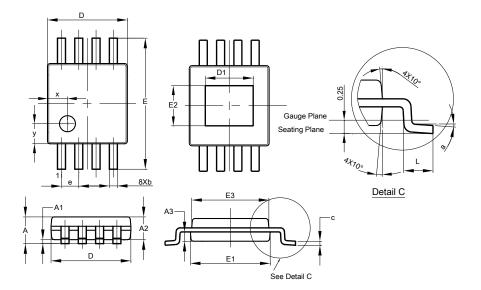
SOT89-5				
Dim	Min	Max	Тур	
Α	1.40	1.60	1.50	
В	0.50	0.62	0.56	
B1	0.44	0.54	0.48	
С	0.35	0.43	0.38	
D	4.40	4.60	4.50	
D1	1.62	1.83	1.733	
Е	2.40	2.60	2.50	
е	-	-	1.50	
Н	3.95	4.25	4.10	
L	0.65	0.95	0.80	
All	All Dimensions in mm			



Package Outline Dimensions (Cont.)

Please see http://www.diodes.com/package-outlines.html for the latest version.

(3) MSOP-8EP

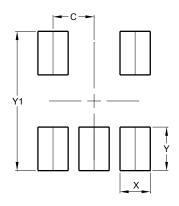


MSOP-8EP				
Dim	Min	Max	Тур	
Α	-	1.10	-	
A1	0.05	0.15	0.10	
A2	0.75	0.95	0.86	
A3	0.29	0.49	0.39	
b	0.22	0.38	0.30	
С	0.08	0.23	0.15	
D	2.90	3.10	3.00	
D1	1.60	2.00	1.80	
E	4.70	5.10	4.90	
E1	2.90	3.10	3.00	
E2	1.30	1.70	1.50	
E3	2.85	3.05	2.95	
е	-	-	0.65	
L	0.40	0.80	0.60	
а	0°	8°	4°	
х	-		0.750	
у	-	-	0.750	
All Dimensions in mm				

Suggested Pad Layout

Please see http://www.diodes.com/package-outlines.html for the latest version.

(1) Package Type: TSOT25



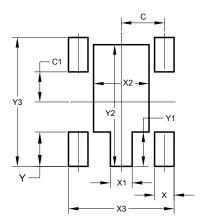
Dimensions	Value (in mm)
С	0.950
Х	0.700
Υ	1.000
Y1	3 199



Suggested Pad Layout (Cont.)

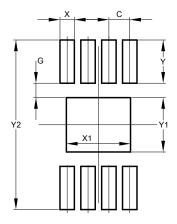
Please see http://www.diodes.com/package-outlines.html for the latest version.

(2) SOT89-5



Dimensions	Value
פווטופוושום	(in mm)
С	1.500
C1	1.050
X	0.680
X1	0.760
X2	1.930
Х3	3.680
Υ	1.200
Y1	1.200
Y2	4.250
Y3	4.500

(3) MSOP-8EP



Dimensions	Value (in mm)
С	0.650
G	0.450
Х	0.450
X1	2.000
Υ	1.350
Y1	1.700
Y2	5 300



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